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(54) **SUCCESSIVE APPROXIMATION
ANALOG-TO-DIGITAL CONVERTER (ADC)
WITH DYNAMIC SEARCH ALGORITHM**

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continuation of application No. 14/248,851, filed on
Apr. 9, 2014, now Pat. No. 8,928,506.

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(2013.01); **H03M 1/04** (2013.01); **H03M 1/44**
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H03M 1/164; H03M 1/201; H03M 1/182;
H03M 1/183; H03M 1/361; H03M 1/0809
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See application file for complete search history.

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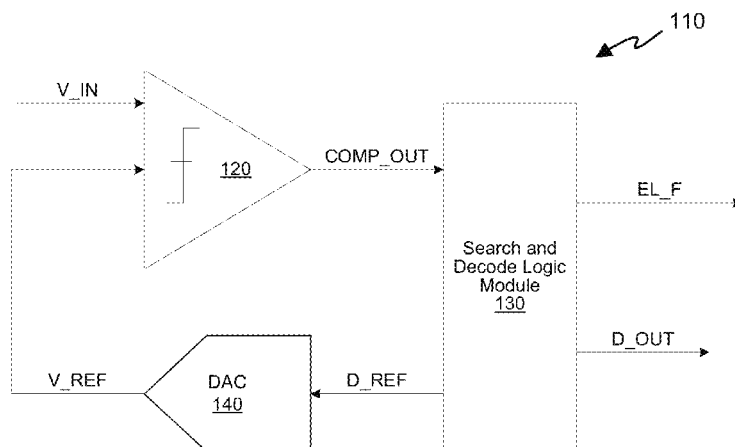
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(57) **ABSTRACT**

Aspects of a method and system for a successive approxi-
mation analog-to-digital converter with dynamic search
algorithms are provided. In some embodiments, a successive
approximation analog-to-digital converter includes a digital-
to-analog converter, a comparator, and a search and decode
logic modules which cooperate to generate a digital output
code representative of the analog input voltage based on a
dynamic search algorithm. The dynamic search algorithms
may alter a sequence of reference voltages used to succes-
sively approximate the analog input voltage based on one or
more characteristics of the analog input voltage.

20 Claims, 18 Drawing Sheets



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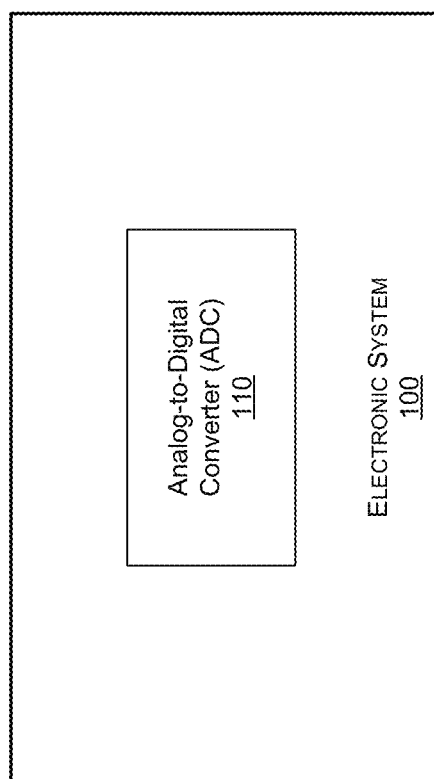


FIG. 1

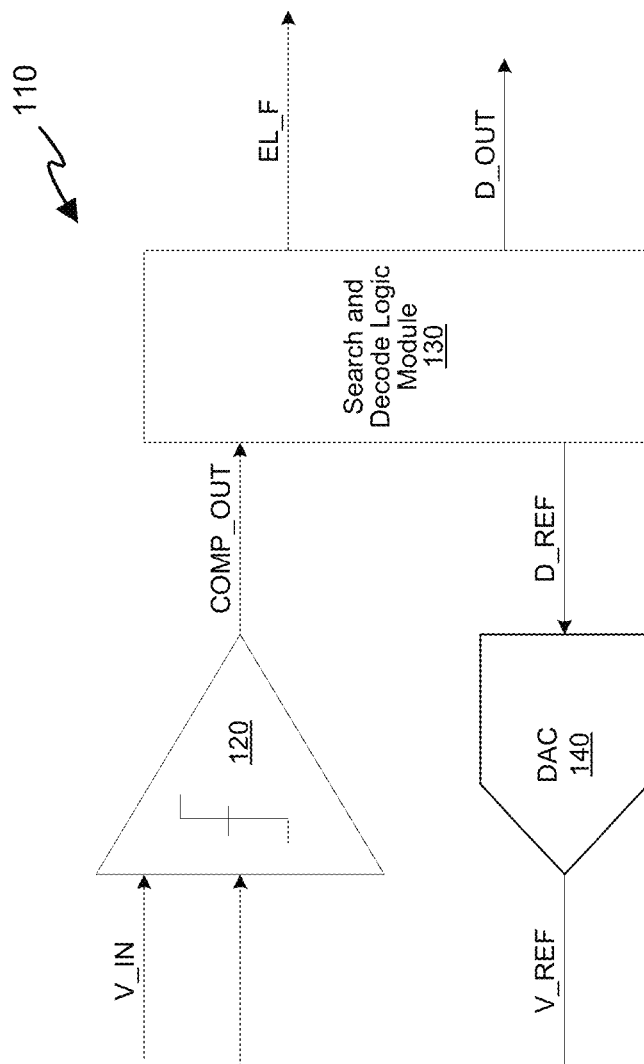


FIG. 2

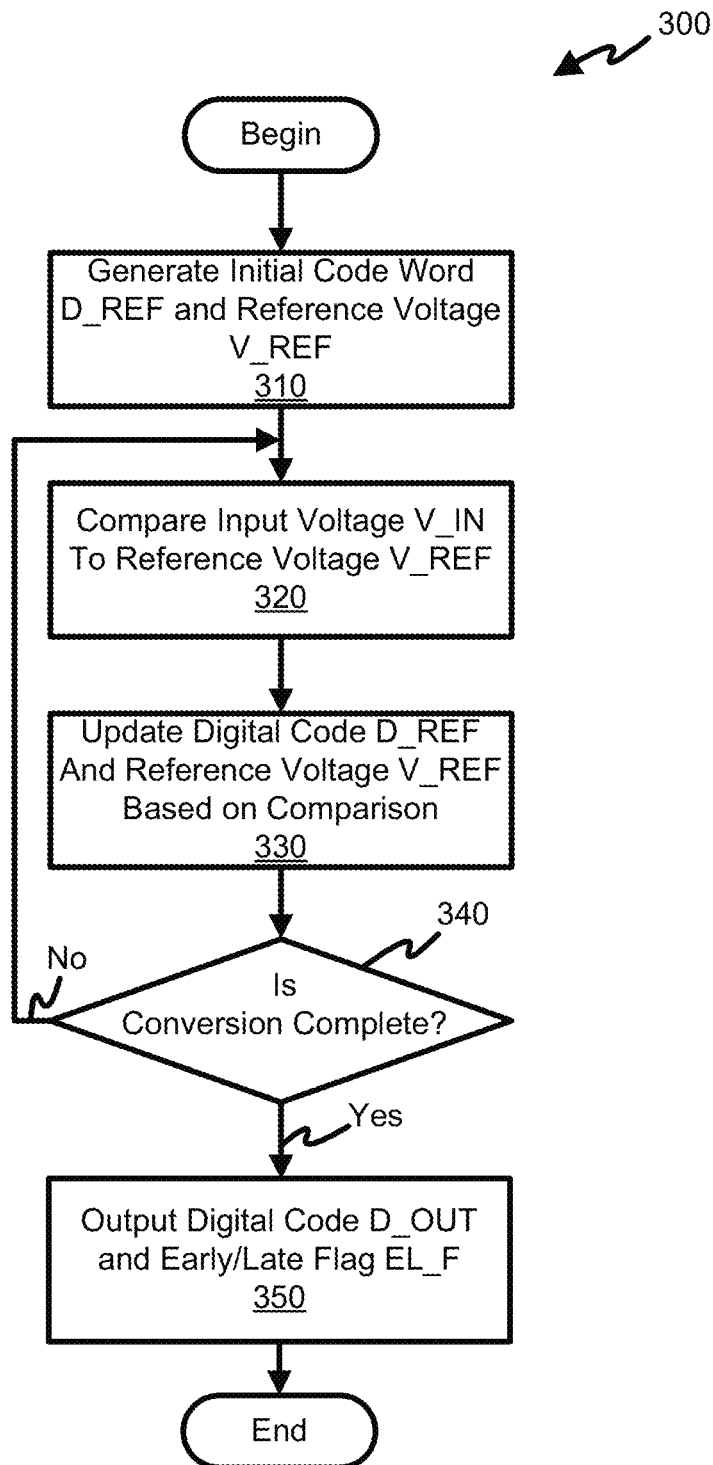


Fig. 3

		7 bits, Vin=7.01						
		Step1	Step2	Step3	Step4	Step5	Step6	Step7
63	0111111		<	<	<	<	<	<
62	0111110		<	<	<	<	<	<
61	0111101		<	<	<	<	<	<
60	0111100		<	<	<	<	<	<
59	0111011		<	<	<	<	<	<
58	0111010		<	<	<	<	<	<
57	0111001		<	<	<	<	<	<
56	0111000		<	<	<	<	<	<
55	0110111		<	<	<	<	<	<
54	0110110		<	<	<	<	<	<
53	0110101		<	<	<	<	<	<
52	0110100		<	<	<	<	<	<
51	0110011		<	<	<	<	<	<
50	0110010		<	<	<	<	<	<
49	0110001		<	<	<	<	<	<
48	0110000		<	<	<	<	<	<
47	0101111		<	<	<	<	<	<
46	0101110		<	<	<	<	<	<
45	0101101		<	<	<	<	<	<
44	0101100		<	<	<	<	<	<
43	0101011		<	<	<	<	<	<
42	0101010		<	<	<	<	<	<
41	0101001		<	<	<	<	<	<
40	0101000		<	<	<	<	<	<
39	0100111		<	<	<	<	<	<
38	0100110		<	<	<	<	<	<
37	0100101		<	<	<	<	<	<
36	0100100		<	<	<	<	<	<
35	0100011		<	<	<	<	<	<
34	0100010		<	<	<	<	<	<
33	0100001		<	<	<	<	<	<
32	0100000			<	<	<	<	<
31	0011111			<	<	<	<	<
30	0011110			<	<	<	<	<
29	0011101			<	<	<	<	<
28	0011100			<	<	<	<	<
27	0011011			<	<	<	<	<
26	0011010			<	<	<	<	<
25	0011001			<	<	<	<	<
24	0011000			<	<	<	<	<
23	0010111			<	<	<	<	<
22	0010110			<	<	<	<	<
21	0010101			<	<	<	<	<
20	0010100			<	<	<	<	<
19	0010011			<	<	<	<	<
18	0010010			<	<	<	<	<
17	0010001			<	<	<	<	<
16	0010000				<	<	<	<
15	0001111				<	<	<	<
14	0001110				<	<	<	<
13	0001101				<	<	<	<
12	0001100				<	<	<	<
11	0001011				<	<	<	<
10	0001010				<	<	<	<
9	0001001				<	<	<	<
8	0001000					<	<	<
7	0000111							
6	0000110						>	>
5	0000101						>	>
4	0000100						>	>
3	0000011					>	>	>
2	0000010					>	>	>
1	0000001					>	>	>
0	0000000		>	>	>	>	>	>

FIG. 4A

7 bits, Vin=15.01								
		Step1	Step2	Step3	Step4	Step5	Step6	Step7
63	0111111		<	<	<	<	<	<
62	0111110		<	<	<	<	<	<
61	0111101		<	<	<	<	<	<
60	0111100		<	<	<	<	<	<
59	0111011		<	<	<	<	<	<
58	0111010		<	<	<	<	<	<
57	0111001		<	<	<	<	<	<
56	0111000		<	<	<	<	<	<
55	0110111		<	<	<	<	<	<
54	0110110		<	<	<	<	<	<
53	0110101		<	<	<	<	<	<
52	0110100		<	<	<	<	<	<
51	0110011		<	<	<	<	<	<
50	0110010		<	<	<	<	<	<
49	0110001		<	<	<	<	<	<
48	0110000		<	<	<	<	<	<
47	0101111		<	<	<	<	<	<
46	0101110		<	<	<	<	<	<
45	0101101		<	<	<	<	<	<
44	0101100		<	<	<	<	<	<
43	0101011		<	<	<	<	<	<
42	0101010		<	<	<	<	<	<
41	0101001		<	<	<	<	<	<
40	0101000		<	<	<	<	<	<
39	0100111		<	<	<	<	<	<
38	0100110		<	<	<	<	<	<
37	0100101		<	<	<	<	<	<
36	0100100		<	<	<	<	<	<
35	0100011		<	<	<	<	<	<
34	0100010		<	<	<	<	<	<
33	0100001		<	<	<	<	<	<
32	0100000			<	<	<	<	<
31	0011111			<	<	<	<	<
30	0011110			<	<	<	<	<
29	0011101			<	<	<	<	<
28	0011100			<	<	<	<	<
27	0011011			<	<	<	<	<
26	0011010			<	<	<	<	<
25	0011001			<	<	<	<	<
24	0011000			<	<	<	<	<
23	0010111			<	<	<	<	<
22	0010110			<	<	<	<	<
21	0010101			<	<	<	<	<
20	0010100			<	<	<	<	<
19	0010011			<	<	<	<	<
18	0010010			<	<	<	<	<
17	0010001			<	<	<	<	<
16	0010000				<	<	<	<
15	0001111							
14	0001110							
13	0001101							
12	0001100							
11	0001011							
10	0001010							
9	0001001							
8	0001000							
7	0000111							
6	0000110							
5	0000101							
4	0000100							
3	0000011							
2	0000010							
1	0000001							
0	0000000							

FIG. 4B

7 bits, Vin=31.01								
		Step1	Step2	Step3	Step4	Step5	Step6	Step7
63	0111111		<	<	<	<	<	<
62	0111110		<	<	<	<	<	<
61	0111101		<	<	<	<	<	<
60	0111100		<	<	<	<	<	<
59	0111011		<	<	<	<	<	<
58	0111010		<	<	<	<	<	<
57	0111001		<	<	<	<	<	<
56	0111000		<	<	<	<	<	<
55	0110111		<	<	<	<	<	<
54	0110110		<	<	<	<	<	<
53	0110101		<	<	<	<	<	<
52	0110100		<	<	<	<	<	<
51	0110011		<	<	<	<	<	<
50	0110010		<	<	<	<	<	<
49	0110001		<	<	<	<	<	<
48	0110000		<	<	<	<	<	<
47	0101111		<	<	<	<	<	<
46	0101110		<	<	<	<	<	<
45	0101101		<	<	<	<	<	<
44	0101100		<	<	<	<	<	<
43	0101011		<	<	<	<	<	<
42	0101010		<	<	<	<	<	<
41	0101001		<	<	<	<	<	<
40	0101000		<	<	<	<	<	<
39	0100111		<	<	<	<	<	<
38	0100110		<	<	<	<	<	<
37	0100101		<	<	<	<	<	<
36	0100100		<	<	<	<	<	<
35	0100011		<	<	<	<	<	<
34	0100010		<	<	<	<	<	<
33	0100001		<	<	<	<	<	<
32	0100000			<	<	<	<	<
31	0011111							
30	0011110							
29	0011101							
28	0011100							
27	0011011							
26	0011010							
25	0011001							
24	0011000							
23	0010111							
22	0010110							
21	0010101							
20	0010100							
19	0010011							
18	0010010							
17	0010001							
16	0010000							
15	0001111							
14	0001110							
13	0001101							
12	0001100							
11	0001011							
10	0001010							
9	0001001							
8	0001000							
7	0000111							
6	0000110							
5	0000101							
4	0000100							
3	0000011							
2	0000010							
1	0000001							
0	0000000							

FIG. 4C

		7 bits, Vin=62.01						
		Step1	Step2	Step3	Step4	Step5	Step6	Step7
63	0111111							
62	0111110							
61	0111101							
60	0111100							
59	0111011							
58	0111010							
57	0111001							
56	0111000							
55	0110111							
54	0110110							
53	0110101							
52	0110100							
51	0110011							
50	0110010							
49	0110001							
48	0110000							
47	0101111							
46	0101110							
45	0101101							
44	0101100							
43	0101011							
42	0101010							
41	0101001							
40	0101000							
39	0100111							
38	0100110							
37	0100101							
36	0100100							
35	0100011							
34	0100010							
33	0100001							
32	0100000							
31	0011111							
30	0011110							
29	0011101							
28	0011100							
27	0011011							
26	0011010							
25	0011001							
24	0011000							
23	0010111							
22	0010110							
21	0010101							
20	0010100							
19	0010011							
18	0010010							
17	0010001							
16	0010000							
15	0001111							
14	0001110							
13	0001101							
12	0001100							
11	0001011							
10	0001010							
9	0001001							
8	0001000							
7	0000111							
6	0000110							
5	0000101							
4	0000100							
3	0000011							
2	0000010							
1	0000001							
0	0000000							

FIG. 4D

7 bits, Vin=7.01							
		Step1	Step2	Step3	Step4	Step5	Step6
63	0111111		<	<	<	<	<
62	0111110		<	<	<	<	<
61	0111101		<	<	<	<	<
60	0111100		<	<	<	<	<
59	0111011		<	<	<	<	<
58	0111010		<	<	<	<	<
57	0111001		<	<	<	<	<
56	0111000		<	<	<	<	<
55	0110111		<	<	<	<	<
54	0110110		<	<	<	<	<
53	0110101		<	<	<	<	<
52	0110100		<	<	<	<	<
51	0110011		<	<	<	<	<
50	0110010		<	<	<	<	<
49	0110001		<	<	<	<	<
48	0110000		<	<	<	<	<
47	0101111		<	<	<	<	<
46	0101110		<	<	<	<	<
45	0101101		<	<	<	<	<
44	0101100		<	<	<	<	<
43	0101011		<	<	<	<	<
42	0101010		<	<	<	<	<
41	0101001		<	<	<	<	<
40	0101000		<	<	<	<	<
39	0100111		<	<	<	<	<
38	0100110		<	<	<	<	<
37	0100101		<	<	<	<	<
36	0100100		<	<	<	<	<
35	0100011		<	<	<	<	<
34	0100010		<	<	<	<	<
33	0100001		<	<	<	<	<
32	0100000		<	<	<	<	<
31	0011111		<	<	<	<	<
30	0011110		<	<	<	<	<
29	0011101		<	<	<	<	<
28	0011100		<	<	<	<	<
27	0011011		<	<	<	<	<
26	0011010		<	<	<	<	<
25	0011001		<	<	<	<	<
24	0011000		<	<	<	<	<
23	0010111		<	<	<	<	<
22	0010110		<	<	<	<	<
21	0010101		<	<	<	<	<
20	0010100		<	<	<	<	<
19	0010011		<	<	<	<	<
18	0010010		<	<	<	<	<
17	0010001		<	<	<	<	<
16	0010000		<	<	<	<	<
15	0001111		<	<	<	<	<
14	0001110		<	<	<	<	<
13	0001101		<	<	<	<	<
12	0001100		<	<	<	<	<
11	0001011		<	<	<	<	<
10	0001010		<	<	<	<	<
9	0001001		<	<	<	<	<
8	0001000		<	<	<	<	<
7	0000111		<	<	<	<	<
6	0000110		<	<	<	<	<
5	0000101		<	<	<	<	<
4	0000100		<	<	<	<	<
3	0000011		<	<	<	<	<
2	0000010		<	<	<	<	<
1	0000001		<	<	<	<	<
0	0000000	<	<	<	<	<	<

FIG. 5A

		7 bits, Vin=15.01				
		Step1	Step2	Step3	Step4	Step5
						Step6
63	0111111		<	<	<	<
62	0111110		<	<	<	<
61	0111101		<	<	<	<
60	0111100		<	<	<	<
59	0111011		<	<	<	<
58	0111010		<	<	<	<
57	0111001		<	<	<	<
56	0111000		<	<	<	<
55	0110111		<	<	<	<
54	0110110		<	<	<	<
53	0110101		<	<	<	<
52	0110100		<	<	<	<
51	0110011		<	<	<	<
50	0110010		<	<	<	<
49	0110001		<	<	<	<
48	0110000		<	<	<	<
47	0101111		<	<	<	<
46	0101110		<	<	<	<
45	0101101		<	<	<	<
44	0101100		<	<	<	<
43	0101011		<	<	<	<
42	0101010		<	<	<	<
41	0101001		<	<	<	<
40	0101000		<	<	<	<
39	0100111		<	<	<	<
38	0100110		<	<	<	<
37	0100101		<	<	<	<
36	0100100		<	<	<	<
35	0100011		<	<	<	<
34	0100010		<	<	<	<
33	0100001		<	<	<	<
32	0100000		<	<	<	<
31	0011111		<	<	<	<
30	0011110		<	<	<	<
29	0011101		<	<	<	<
28	0011100		<	<	<	<
27	0011011		<	<	<	<
26	0011010		<	<	<	<
25	0011001		<	<	<	<
24	0011000		<	<	<	<
23	0010111		<	<	<	<
22	0010110		<	<	<	<
21	0010101		<	<	<	<
20	0010100		<	<	<	<
19	0010011		<	<	<	<
18	0010010		<	<	<	<
17	0010001		<	<	<	<
16	0010000		<	<	<	<
15	0001111		<	<	<	<
14	0001110		<	<	<	<
13	0001101		<	<	<	<
12	0001100		<	<	<	<
11	0001011		<	<	<	<
10	0001010		<	<	<	<
9	0001001		<	<	<	<
8	0001000		<	<	<	<
7	0000111		<	<	<	<
6	0000110		<	<	<	<
5	0000101		<	<	<	<
4	0000100		<	<	<	<
3	0000011		<	<	<	<
2	0000010		<	<	<	<
1	0000001		<	<	<	<
0	0000000		<	<	<	<

FIG. 5B

		7 bits, Vin=31.01						
		Step1	Step2	Step3	Step4	Step5	Step6	Step7
63	0111111			<	<	<	<	<
62	0111110			<	<	<	<	<
61	0111101			<	<	<	<	<
60	0111100			<	<	<	<	<
59	0111011			<	<	<	<	<
58	0111010			<	<	<	<	<
57	0111001			<	<	<	<	<
56	0111000			<	<	<	<	<
55	0110111			<	<	<	<	<
54	0110110			<	<	<	<	<
53	0110101			<	<	<	<	<
52	0110100			<	<	<	<	<
51	0110011			<	<	<	<	<
50	0110010			<	<	<	<	<
49	0110001			<	<	<	<	<
48	0110000			<	<	<	<	<
47	0101111			<	<	<	<	<
46	0101110			<	<	<	<	<
45	0101101			<	<	<	<	<
44	0101100			<	<	<	<	<
43	0101011			<	<	<	<	<
42	0101010			<	<	<	<	<
41	0101001			<	<	<	<	<
40	0101000			<	<	<	<	<
39	0100111			<	<	<	<	<
38	0100110			<	<	<	<	<
37	0100101			<	<	<	<	<
36	0100100			<	<	<	<	<
35	0100011			<	<	<	<	<
34	0100010			<	<	<	<	<
33	0100001			<	<	<	<	<
32	0100000			<	<	<	<	<
31	0011111							
30	0011110							
29	0011101							
28	0011100							
27	0011011							
26	0011010							
25	0011001							
24	0011000							
23	0010111							
22	0010110							
21	0010101							
20	0010100							
19	0010011							
18	0010010							
17	0010001							
16	0010000							
15	0001111							
14	0001110							
13	0001101							
12	0001100							
11	0001011							
10	0001010							
9	0001001							
8	0001000							
7	0000111							
6	0000110							
5	0000101							
4	0000100							
3	0000011							
2	0000010							
1	0000001							
0	0000000							

FIG. 5C

		7 bits, Vin=62.01							
		Step1	Step2	Step3	Step4	Step5	Step6	Step7	Step8
63	0111111								*
62	0111110							*	>
61	0111101							>	>
60	0111100						*	>	>
59	0111011						>	>	>
58	0111010						>	>	>
57	0111001						>	>	>
56	0111000					*	>	>	>
55	0110111					>	>	>	>
54	0110110					>	>	>	>
53	0110101					>	>	>	>
52	0110100					>	>	>	>
51	0110011					>	>	>	>
50	0110010					>	>	>	>
49	0110001					>	>	>	>
48	0110000				*	>	>	>	>
47	0101111				>	>	>	>	>
46	0101110				>	>	>	>	>
45	0101101				>	>	>	>	>
44	0101100				>	>	>	>	>
43	0101011				>	>	>	>	>
42	0101010				>	>	>	>	>
41	0101001				>	>	>	>	>
40	0101000				>	>	>	>	>
39	0100111				>	>	>	>	>
38	0100110				>	>	>	>	>
37	0100101				>	>	>	>	>
36	0100100				>	>	>	>	>
35	0100011				>	>	>	>	>
34	0100010				>	>	>	>	>
33	0100001				>	>	>	>	>
32	0100000			*	>	>	>	>	>
31	0011111			>	>	>	>	>	>
30	0011110			>	>	>	>	>	>
29	0011101			>	>	>	>	>	>
28	0011100			>	>	>	>	>	>
27	0011011			>	>	>	>	>	>
26	0011010			>	>	>	>	>	>
25	0011001			>	>	>	>	>	>
24	0011000			>	>	>	>	>	>
23	0010111			>	>	>	>	>	>
22	0010110			>	>	>	>	>	>
21	0010101			>	>	>	>	>	>
20	0010100			>	>	>	>	>	>
19	0010011			>	>	>	>	>	>
18	0010010			>	>	>	>	>	>
17	0010001			>	>	>	>	>	>
16	0010000		*	>	>	>	>	>	>
15	0001111		>	>	>	>	>	>	>
14	0001110		>	>	>	>	>	>	>
13	0001101		>	>	>	>	>	>	>
12	0001100		>	>	>	>	>	>	>
11	0001011		>	>	>	>	>	>	>
10	0001010		>	>	>	>	>	>	>
9	0001001		>	>	>	>	>	>	>
8	0001000		>	>	>	>	>	>	>
7	0000111		>	>	>	>	>	>	>
6	0000110		>	>	>	>	>	>	>
5	0000101		>	>	>	>	>	>	>
4	0000100		>	>	>	>	>	>	>
3	0000011		>	>	>	>	>	>	>
2	0000010		>	>	>	>	>	>	>
1	0000001		>	>	>	>	>	>	>
0	0000000	*	>	>	>	>	>	>	>

FIG. 5D

		7 bits, Vin=7.01				
		Step1	Step2	Step3	Step4	Step5
63	0111111		<	<	<	<
62	0111110		<	<	<	<
61	0111101		<	<	<	<
60	0111100		<	<	<	<
59	0111011		<	<	<	<
58	0111010		<	<	<	<
57	0111001		<	<	<	<
56	0111000		<	<	<	<
55	0110111		<	<	<	<
54	0110110		<	<	<	<
53	0110101		<	<	<	<
52	0110100		<	<	<	<
51	0110011		<	<	<	<
50	0110010		<	<	<	<
49	0110001		<	<	<	<
48	0110000		<	<	<	<
47	0101111		<	<	<	<
46	0101110		<	<	<	<
45	0101101		<	<	<	<
44	0101100		<	<	<	<
43	0101011		<	<	<	<
42	0101010		<	<	<	<
41	0101001		<	<	<	<
40	0101000		<	<	<	<
39	0100111		<	<	<	<
38	0100110		<	<	<	<
37	0100101		<	<	<	<
36	0100100		<	<	<	<
35	0100011		<	<	<	<
34	0100010		<	<	<	<
33	0100001		<	<	<	<
32	0100000		<	<	<	<
31	0011111		<	<	<	<
30	0011110		<	<	<	<
29	0011101		<	<	<	<
28	0011100		<	<	<	<
27	0011011		<	<	<	<
26	0011010		<	<	<	<
25	0011001		<	<	<	<
24	0011000		<	<	<	<
23	0010111		<	<	<	<
22	0010110		<	<	<	<
21	0010101		<	<	<	<
20	0010100		<	<	<	<
19	0010011		<	<	<	<
18	0010010		<	<	<	<
17	0010001		<	<	<	<
16	0010000		<	<	<	<
15	0001111		<	<	<	<
14	0001110		<	<	<	<
13	0001101		<	<	<	<
12	0001100		<	<	<	<
11	0001011		<	<	<	<
10	0001010		<	<	<	<
9	0001001		<	<	<	<
8	0001000		<	<	<	<
7	0000111				<	<
6	0000110				<	<
5	0000101				<	<
4	0000100				<	<
3	0000011				<	<
2	0000010				<	<
1	0000001				<	<
0	0000000				<	<

FIG. 6A

7 bits, Vin=15.01						
	Step1	Step2	Step3	Step4	Step5	Step6 Step7
63	0111111		<	<	<	<
62	0111110		<	<	<	<
61	0111101		<	<	<	<
60	0111100		<	<	<	<
59	0111011		<	<	<	<
58	0111010		<	<	<	<
57	0111001		<	<	<	<
56	0111000		<	<	<	<
55	0110111		<	<	<	<
54	0110110		<	<	<	<
53	0110101		<	<	<	<
52	0110100		<	<	<	<
51	0110011		<	<	<	<
50	0110010		<	<	<	<
49	0110001		<	<	<	<
48	0110000		<	<	<	<
47	0101111		<	<	<	<
46	0101110		<	<	<	<
45	0101101		<	<	<	<
44	0101100		<	<	<	<
43	0101011		<	<	<	<
42	0101010		<	<	<	<
41	0101001		<	<	<	<
40	0101000		<	<	<	<
39	0100111		<	<	<	<
38	0100110		<	<	<	<
37	0100101		<	<	<	<
36	0100100		<	<	<	<
35	0100011		<	<	<	<
34	0100010		<	<	<	<
33	0100001		<	<	<	<
32	0100000		<	<	<	<
31	0011111		<	<	<	<
30	0011110		<	<	<	<
29	0011101		<	<	<	<
28	0011100		<	<	<	<
27	0011011		<	<	<	<
26	0011010		<	<	<	<
25	0011001		<	<	<	<
24	0011000		<	<	<	<
23	0010111		<	<	<	<
22	0010110		<	<	<	<
21	0010101		<	<	<	<
20	0010100		<	<	<	<
19	0010011		<	<	<	<
18	0010010		<	<	<	<
17	0010001		<	<	<	<
16	0010000		<	<	<	<
15	0001111		<	<	<	<
14	0001110		<	<	<	<
13	0001101		<	<	<	<
12	0001100		<	<	<	<
11	0001011		<	<	<	<
10	0001010		<	<	<	<
9	0001001		<	<	<	<
8	0001000		<	<	<	<
7	0000111		<	<	<	<
6	0000110		<	<	<	<
5	0000101		<	<	<	<
4	0000100		<	<	<	<
3	0000011		<	<	<	<
2	0000010		<	<	<	<
1	0000001		<	<	<	<
0	0000000		<	<	<	<

FIG. 6B

7 bits, Vin=31.01									
		Step1	Step2	Step3	Step4	Step5	Step6	Step7	Step8
63	0111111			<	<	<	<	<	<
62	0111110			<	<	<	<	<	<
61	0111101			<	<	<	<	<	<
60	0111100			<	<	<	<	<	<
59	0111011			<	<	<	<	<	<
58	0111010			<	<	<	<	<	<
57	0111001			<	<	<	<	<	<
56	0111000			<	<	<	<	<	<
55	0110111			<	<	<	<	<	<
54	0110110			<	<	<	<	<	<
53	0110101			<	<	<	<	<	<
52	0110100			<	<	<	<	<	<
51	0110011			<	<	<	<	<	<
50	0110010			<	<	<	<	<	<
49	0110001			<	<	<	<	<	<
48	0110000			<	<	<	<	<	<
47	0101111			<	<	<	<	<	<
46	0101110			<	<	<	<	<	<
45	0101101			<	<	<	<	<	<
44	0101100			<	<	<	<	<	<
43	0101011			<	<	<	<	<	<
42	0101010			<	<	<	<	<	<
41	0101001			<	<	<	<	<	<
40	0101000			<	<	<	<	<	<
39	0100111			<	<	<	<	<	<
38	0100110			<	<	<	<	<	<
37	0100101			<	<	<	<	<	<
36	0100100			<	<	<	<	<	<
35	0100011			<	<	<	<	<	<
34	0100010			<	<	<	<	<	<
33	0100001			<	<	<	<	<	<
32	0100000			<	<	<	<	<	<
31	0011111								
30	0011110								
29	0011101								
28	0011100								
27	0011011								
26	0011010								
25	0011001								
24	0011000								
23	0010111								
22	0010110								
21	0010101								
20	0010100								
19	0010011								
18	0010010								
17	0010001								
16	0010000								
15	0001111								
14	0001110								
13	0001101								
12	0001100								
11	0001011								
10	0001010								
9	0001001								
8	0001000								
7	0000111								
6	0000110								
5	0000101								
4	0000100								
3	0000011								
2	0000010								
1	0000001								
0	0000000								

FIG. 6C

		7 bits, Vin=62.01							
		Step1	Step2	Step3	Step4	Step5	Step6	Step7	Step8
63	0111111								
62	0111110							*	>
61	0111101							>	>
60	0111100						*	>	>
59	0111011						>	>	>
58	0111010						>	>	>
57	0111001						>	>	>
56	0111000					*	>	>	>
55	0110111					>	>	>	>
54	0110110					>	>	>	>
53	0110101					>	>	>	>
52	0110100					>	>	>	>
51	0110011					>	>	>	>
50	0110010					>	>	>	>
49	0110001					>	>	>	>
48	0110000				*	>	>	>	>
47	0101111				>	>	>	>	>
46	0101110				>	>	>	>	>
45	0101101				>	>	>	>	>
44	0101100				>	>	>	>	>
43	0101011				>	>	>	>	>
42	0101010				>	>	>	>	>
41	0101001				>	>	>	>	>
40	0101000				>	>	>	>	>
39	0100111				>	>	>	>	>
38	0100110				>	>	>	>	>
37	0100101				>	>	>	>	>
36	0100100				>	>	>	>	>
35	0100011				>	>	>	>	>
34	0100010				>	>	>	>	>
33	0100001				>	>	>	>	>
32	0100000			*	>	>	>	>	>
31	0011111			>	>	>	>	>	>
30	0011110			>	>	>	>	>	>
29	0011101			>	>	>	>	>	>
28	0011100			>	>	>	>	>	>
27	0011011			>	>	>	>	>	>
26	0011010			>	>	>	>	>	>
25	0011001			>	>	>	>	>	>
24	0011000			>	>	>	>	>	>
23	0010111			>	>	>	>	>	>
22	0010110			>	>	>	>	>	>
21	0010101			>	>	>	>	>	>
20	0010100			>	>	>	>	>	>
19	0010011			>	>	>	>	>	>
18	0010010			>	>	>	>	>	>
17	0010001			>	>	>	>	>	>
16	0010000			>	>	>	>	>	>
15	0001111			>	>	>	>	>	>
14	0001110			>	>	>	>	>	>
13	0001101			>	>	>	>	>	>
12	0001100			>	>	>	>	>	>
11	0001011			>	>	>	>	>	>
10	0001010			>	>	>	>	>	>
9	0001001			>	>	>	>	>	>
8	0001000		*	>	>	>	>	>	>
7	0000111		>	>	>	>	>	>	>
6	0000110		>	>	>	>	>	>	>
5	0000101		>	>	>	>	>	>	>
4	0000100		>	>	>	>	>	>	>
3	0000011		>	>	>	>	>	>	>
2	0000010		>	>	>	>	>	>	>
1	0000001		>	>	>	>	>	>	>
0	0000000	*	>	>	>	>	>	>	>

FIG. 6D

		7 bits, Vin=15.01					
		Step1	Step2	Step3	Step4	Step5	Step6
63	0111111			<	<	<	<
62	0111110			<	<	<	<
61	0111101			<	<	<	<
60	0111100			<	<	<	<
59	0111011			<	<	<	<
58	0111010			<	<	<	<
57	0111001			<	<	<	<
56	0111000			<	<	<	<
55	0110111			<	<	<	<
54	0110110			<	<	<	<
53	0110101			<	<	<	<
52	0110100			<	<	<	<
51	0110011			<	<	<	<
50	0110010			<	<	<	<
49	0110001			<	<	<	<
48	0110000			<	<	<	<
47	0101111			<	<	<	<
46	0101110			<	<	<	<
45	0101101			<	<	<	<
44	0101100			<	<	<	<
43	0101011			<	<	<	<
42	0101010			<	<	<	<
41	0101001			<	<	<	<
40	0101000			<	<	<	<
39	0100111			<	<	<	<
38	0100110			<	<	<	<
37	0100101			<	<	<	<
36	0100100			<	<	<	<
35	0100011			<	<	<	<
34	0100010			<	<	<	<
33	0100001			<	<	<	<
32	0100000			<	<	<	<
31	0011111			<	<	<	<
30	0011110			<	<	<	<
29	0011101			<	<	<	<
28	0011100			<	<	<	<
27	0011011			<	<	<	<
26	0011010			<	<	<	<
25	0011001			<	<	<	<
24	0011000			<	<	<	<
23	0010111			<	<	<	<
22	0010110			<	<	<	<
21	0010101			<	<	<	<
20	0010100			<	<	<	<
19	0010011			<	<	<	<
18	0010010			<	<	<	<
17	0010001			<	<	<	<
16	0010000			<	<	<	<
15	0001111						>
14	0001110						>
13	0001101						>
12	0001100						>
11	0001011						>
10	0001010						>
9	0001001						>
8	0001000						>
7	0000111						>
6	0000110						>
5	0000101						>
4	0000100						>
3	0000011						>
2	0000010						>
1	0000001						>
0	0000000						>

FIG. 7A

7 bits, Vin=31.01									
		Step1	Step2	Step3	Step4	Step5	Step6	Step7	Step8
63	0111111				<	<	<	<	<
62	0111110				<	<	<	<	<
61	0111101				<	<	<	<	<
60	0111100				<	<	<	<	<
59	0111011				<	<	<	<	<
58	0111010				<	<	<	<	<
57	0111001				<	<	<	<	<
56	0111000				<	<	<	<	<
55	0110111				<	<	<	<	<
54	0110110				<	<	<	<	<
53	0110101				<	<	<	<	<
52	0110100				<	<	<	<	<
51	0110011				<	<	<	<	<
50	0110010				<	<	<	<	<
49	0110001				<	<	<	<	<
48	0110000				<	<	<	<	<
47	0101111				<	<	<	<	<
46	0101110				<	<	<	<	<
45	0101101				<	<	<	<	<
44	0101100				<	<	<	<	<
43	0101011				<	<	<	<	<
42	0101010				<	<	<	<	<
41	0101001				<	<	<	<	<
40	0101000				<	<	<	<	<
39	0100111				<	<	<	<	<
38	0100110				<	<	<	<	<
37	0100101				<	<	<	<	<
36	0100100				<	<	<	<	<
35	0100011				<	<	<	<	<
34	0100010				<	<	<	<	<
33	0100001				<	<	<	<	<
32	0100000				<	<	<	<	<
31	0011111				<	<	<	<	<
30	0011110							<	<
29	0011101							<	<
28	0011100							<	<
27	0011011							<	<
26	0011010							<	<
25	0011001							<	<
24	0011000							<	<
23	0010111							<	<
22	0010110							<	<
21	0010101							<	<
20	0010100							<	<
19	0010011							<	<
18	0010010							<	<
17	0010001							<	<
16	0010000							<	<
15	0001111							<	<
14	0001110							<	<
13	0001101							<	<
12	0001100							<	<
11	0001011							<	<
10	0001010							<	<
9	0001001							<	<
8	0001000							<	<
7	0000111							<	<
6	0000110							<	<
5	0000101							<	<
4	0000100							<	<
3	0000011							<	<
2	0000010							<	<
1	0000001							<	<
0	0000000							<	<

FIG. 7B

7 bits, Vin=62.01										
		Step1	Step2	Step3	Step4	Step5	Step6	Step7	Step8	Step9
63	0111111									
62	0111110									
61	0111101									
60	0111100									
59	0111011									
58	0111010									
57	0111001									
56	0111000									
55	0110111									
54	0110110									
53	0110101									
52	0110100									
51	0110011									
50	0110010									
49	0110001									
48	0110000									
47	0101111									
46	0101110									
45	0101101									
44	0101100									
43	0101011									
42	0101010									
41	0101001									
40	0101000									
39	0100111									
38	0100110									
37	0100101									
36	0100100									
35	0100011									
34	0100010									
33	0100001									
32	0100000									
31	0011111									
30	0011110									
29	0011101									
28	0011100									
27	0011011									
26	0011010									
25	0011001									
24	0011000									
23	0010111									
22	0010110									
21	0010101									
20	0010100									
19	0010011									
18	0010010									
17	0010001									
16	0010000									
15	0001111									
14	0001110									
13	0001101									
12	0001100									
11	0001011									
10	0001010									
9	0001001									
8	0001000									
7	0000111									
6	0000110									
5	0000101									
4	0000100									
3	0000011									
2	0000010									
1	0000001									
0	0000000									

FIG. 7C

SUCCESSIVE APPROXIMATION ANALOG-TO-DIGITAL CONVERTER (ADC) WITH DYNAMIC SEARCH ALGORITHM

CLAIM OF PRIORITY

This patent application is a continuation of U.S. patent application Ser. No. 14/558,004, filed Dec. 2, 2014, which is a continuation of U.S. patent application Ser. No. 14/248,851, filed Apr. 9, 2014, which makes reference to, claims priority to and claims benefit from the U.S. Provisional Patent Application Ser. No. 61/809,926, filed on Apr. 9, 2013. The above stated applications are hereby incorporated herein by reference in their entirety.

TECHNICAL FIELD

Certain embodiments of the invention relate to analog-to-digital converters (ADCs). More specifically, certain embodiments of the invention relate to methods and systems successive approximation ADCs and searches performed by successive approximation ADCs.

BACKGROUND

Existing analog-to-digital converters (ADCs) can be too slow and/or power hungry. Further limitations and disadvantages of conventional and traditional approaches will become apparent to one of skill in the art, through comparison of such systems with some aspects of the present invention as set forth in the remainder of the present application with reference to the drawings.

BRIEF SUMMARY OF THE INVENTION

Systems and/or methods are provided for a successive approximation analog-to-digital converter (ADC) with a dynamic search algorithm, substantially as shown in and/or described in connection with at least one of the figures, as set forth more completely in the claims.

These and other advantages, aspects and novel features of the present invention, as well as details of an illustrated embodiment thereof, will be more fully understood from the following description and drawings.

BRIEF DESCRIPTION OF SEVERAL VIEWS OF THE DRAWINGS

FIG. 1 illustrates an example electronic system that may be operable to perform analog-to-digital conversions.

FIG. 2 is a diagram illustrating an example successive approximation ADC.

FIG. 3 is a flow chart illustrating an example process used by the successive approximation ADC of FIG. 2 to obtain a digital code word representative of an analog input voltage.

FIG. 4A-4D illustrates an example search that may be performed by the successive approximation ADC of FIG. 2 per the process of the FIG. 3.

FIG. 5A-5D illustrates another example search that may be performed by the successive approximation ADC of FIG. 2 per the process of the FIG. 3.

FIG. 6A-6D illustrates yet another example search that may be performed by the successive approximation ADC of FIG. 2 per the process of the FIG. 3.

FIG. 7A-7C illustrates a further example search that may be performed by the successive approximation ADC of FIG. 2 per the process of the FIG. 3.

DETAILED DESCRIPTION

As utilized herein the terms “circuits” and “circuitry” refer to physical electronic components (i.e. hardware) and any software and/or firmware (“code”) which may configure the hardware, be executed by the hardware, and/or otherwise be associated with the hardware. As used herein, for example, a particular processor and memory may comprise a first “circuit” when executing a first one or more lines of code and may comprise a second “circuit” when executing a second one or more lines of code. As utilized herein, “and/or” means any one or more of the items in the list joined by “and/or”. As an example, “x and/or y” means any element of the three-element set $\{(x), (y), (x, y)\}$. As another example, “x, y, and/or z” means any element of the seven-element set $\{(x), (y), (z), (x, y), (x, z), (y, z), (x, y, z)\}$. As utilized herein, the term “exemplary” means serving as a non-limiting example, instance, or illustration. As utilized herein, the terms “e.g.,” and “for example” set off lists of one or more non-limiting examples, instances, or illustrations. As utilized herein, circuitry is “operable” to perform a function whenever the circuitry comprises the necessary hardware and code (if any is necessary) to perform the function, regardless of whether performance of the function is disabled, or not enabled, by some user-configurable setting.

Aspects of the invention include an analog-to-digital converter (ADC) customized for use in systems, such as communication systems, where much of the dynamic range of the ADC is present only to avoid clipping in the event of infrequent peaks (e.g., resulting from blocker signals). Because the signals whose amplitudes range into the higher digital codes are typically unwanted blocker signals, larger amounts of quantization noise may be tolerable at such signal levels.

FIG. 1 illustrates an example electronic system 100 that may be operable to perform analog-to-digital conversion. The electronic system 100 may comprise suitable circuitry, interfaces, logic and/or code for implementing various aspects of the present disclosure. In this regard, the electronic system 100 may be configured to support performing, executing or running various operations, functions, applications and/or services. For example, the electronic system 100 may be used for executing computer programs, playing video and/or audio content, gaming, communication applications or services (e.g., Internet access/browsing, email, text messaging, chatting and/or voice calling services), and/or networking services (e.g., WiFi hotspot, Bluetooth piconet, and/or active 3G/femtocell data channels).

In some instances, the electronic system 100 may be configured to enable and/or support communication of data. In this regard, the electronic system 100 may communicate with other systems (local or remote), such as during executing, running, and/or performing of operations, functions, applications and/or services supported by the electronic system 100. For example, the electronic system 100 may be configured to support (e.g., using suitable dedicated communication components or subsystems) use of wired and/or wireless connections/interfaces, which may be configured in accordance with one or more supported wireless and/or wired protocols or standards, to facilitate transmission and/or reception of signals (carrying data) to and/or from the electronic system 100. In this regard, the electronic system 100 may be operable to process transmitted or received signals in accordance with applicable wired or wireless protocols.

Examples of wireless protocols or standards that may be supported and/or used by the communication subsystem **250** may comprise wireless personal area network (WPAN) protocols, such as Bluetooth (IEEE 802.15); near field communication (NFC) standards; wireless local area network (WLAN) protocols, such as WiFi (IEEE 802.11); cellular standards, such as 2G/2G+ (e.g., GSM/GPRS/EDGE, and IS-95 or cdmaOne) and/or 2G/2G+ (e.g., CDMA2000, UMTS, and HSPA); 4G standards, such as WiMAX (IEEE 802.16) and LTE; Ultra-Wideband (UWB), direct broadcast satellite (DBS), microwave backhaul, and/or the like. Examples of wired protocols and/or interfaces that may be supported and/or used by the communication subsystem **250** comprise Ethernet (IEEE 802.2), Fiber Distributed Data Interface (FDDI), Integrated Services Digital Network (ISDN), cable (DOCSIS) and Universal Serial Bus (USB) based interfaces. Examples of signal processing operations that may be performed by the electronic system **100** comprise, for example, filtering, amplification, analog-to-digital conversion and/or digital-to-analog conversion, up-conversion/down-conversion of baseband signals, encoding/decoding, encryption/decryption, and/or modulation/demodulation.

In some instances, the electronic system **100** may be configured to enable or support input/output operations, such as to allow user interactions that may control services provided by the electronic system **100**. In this regard, the electronic system **100** may comprise components or subsystems for enabling interactions with a user (e.g., end-user or installer), so as to obtain user input and/or to provide user output.

The electronic system **100** may be a stationary system (i.e., being installed at, and/or configured for use only in particular location). In other instances, however, the electronic system **100** may be a mobile device—i.e., intended for use on the move and/or at different locations. In this regard, the electronic system **100** may be designed and/or configured (e.g., as handheld device) to allow for ease of movement, such as to allow it to be readily moved while being held by the user as the user moves, and the electronic system **100** may be configured to perform at least some of the operations, functions, applications and/or services supported on the move.

Examples of electronic systems may comprise handheld electronic devices (e.g., cellular phones, smartphones, or tablets), personal computers (e.g., laptops or desktops), servers, dedicated multimedia devices (e.g., televisions, game consoles, or portable media players), DBS outdoor units, DBS indoor units (e.g., “set-top boxes”), microwave backhaul outdoor units, microwave backhaul indoor units, or other similar receiver systems, and the like. The disclosure, however, is not limited to any particular type of electronic system.

In operation, the electronic system **100** may be operable to perform various operations, functions, applications and/or services. In this regard, in some instances, performing the various, operations, functions, application or services supported by the electronic system **100** may entail performing various processing operations on data handled by the electronic system **100**. For example, communication of data, whether over wired or wireless interfaces, may typically comprise transmitting and/or receiving analog signals that are communicated over wireless and/or wired connections. In this regard, typically analog radio frequency (RF) signals may be used to carry data (e.g., content), which may be embedded into the analog signals using analog or digital modulation schemes. For analog communications, data is

transferred using continuously varying analog signals, and for digital communications, the analog signals are used to transfer discrete messages in accordance with a particular digitalization scheme.

Accordingly, handling of digital communications may typically require performing, among other things, analog-to-digital conversion at the receiving end. In this regard, the system **100** may include a successive approximation ADC **110**. The ADC **110** may comprise circuitry, interfaces, logic and/or code for performing analog-to-digital conversions. To this end, the ADC **110** may use a dynamic algorithm for choosing analog reference voltages against which to compare an analog input voltage to be converted to a digital code. The search algorithm may be dynamic in that, for example, a sequence of reference voltages used for a conversion may be selected based on characteristics of the current input signal being converted and/or past converted input signals. The search algorithm may be dynamic in that, for example, a sequence of reference voltages used for a conversion may change from one conversion (or set of conversions) to the next.

In an example implementation, the sequence of reference voltages may be selected based on characteristics of the signal to be converted. Characteristics of the signal may include, for example, signal to noise ratio (SNR), peak-to-average power ratio (PAPR), and/or the like. Characteristics of the signal may include, for example, a cumulative distribution function (CDF) or histogram of the signal being digitized. The sequence of reference voltages may also be selected based on any other suitable factors such as, for example, process variations, voltage variations, temperature variations, and/or the like. Furthermore, the sequence of reference voltages may be selected according to a search algorithm such as, for example, a binary search algorithm, a non-binary search algorithm, and a partial binary search algorithm.

Referring now to FIG. 2, a block diagram is shown that illustrates an example ADC **110**, in connection with various exemplary aspects of the invention. As shown, the ADC **110** may include a comparator **120**, a search and decode logic module **130**, and a digital-to-analog converter (DAC) **140**.

The comparator **120** may comprise suitable logic, circuitry, interfaces and/or code that may be operable to generate an output signal or code COMP_OUT based on analog input voltage V_IN and an analog reference voltage V_REF generated by the DAC **140**. In this regard, the comparator **120** may be operable to generate the output signal COMP_OUT such that it indicates whether the analog input voltage V_IN has a predetermined relationship (e.g., greater than) to the analog reference voltage V_REF generated by the DAC **140**. To this end, the comparator **120** may comprise an analog voltage comparator.

The search and decode logic module **130** may comprise suitable logic, circuitry, interfaces and/or code that may be operable to perform a search of the bits that form a digital reference voltage code D_REF based on the output signal COMP_OUT from the comparator **120**. Moreover, the search and decode logic module **130** may provide the generated digital reference voltage code D_REF to the DAC **140**. In this regard, the search and decode logic module **130** may be operable to search for a reference voltage code D_REF that causes the DAC **140** to produce an analog reference voltage V_REF that approximates the analog input voltage V_IN. To this end, the search and decode logic module **130** may comprise search and decode circuitry that iterates or searches through the range of possible digital reference voltage codes D_REF in accordance with a par-

ticular search algorithm and/or particular search criteria. The manner by which the search and decode logic module 130 may iterate through the range of possible digital reference codes D_REF is explained in greater detail below.

Besides generating the digital reference voltage code D_REF, the search and decode logic module 130 may further generate and output an early/late flag EL_F that indicates whether the search and decode logic module 130 obtained a digital output code D_OUT that is representative of the analog input voltage V_IN early or late with respect to a base line number of steps. In particular, the search and decode logic module 130 may generate the early/late flag EL_F such that it indicates, for example, that the digital output code D_OUT was obtained 1 step early, 1 step late, 2 steps early, 2 steps late, or some other indication.

The DAC 140 may comprise suitable logic, circuitry, interfaces and/or code that may be operable to generate the analog reference voltage V_REF, based on a digital reference voltage code D_REF received from the search and decode logic module 130. As shown, the DAC 140 may communicate the generated analog reference voltage V_REF to the comparator 120.

Referring now to FIG. 3, an example method 300 is depicted that may be implemented by the ADC 110 to generate a digital output code D_OUT that is representative of an analog input voltage V_IN. At 310, the search and decode logic module 130 may cause the DAC 140 to generate an initial reference voltage V_REF. To this end, the search and decode logic module 130 may generate an initial digital reference voltage code D_REF for the DAC 140 which causes the DAC 140 to generate a corresponding analog reference voltage V_REF. At 320, the comparator 120 may compare the analog input voltage V_IN with the analog reference voltage V_REF produced by the DAC 140. Based on such comparison, the comparator 120 may provide the search and decode logic module 130 with a comparison output signal COMP_OUT that is indicative of whether the analog input voltage V_IN has a predetermined relationship (e.g., greater than) to the analog reference voltage V_REF.

At 330, the search and decode logic module 130 may update the analog reference voltage V_REF based upon the comparison output signal COMP_OUT received from the comparator 120 and various search criteria discussed below. To this end, the search and decode logic module 130 may update the digital reference voltage code D_REF that is supplied to the DAC 140 based on the comparison output signal COMP_OUT and search criteria, thus causing the DAC 140 to generate a correspond analog reference voltage V_REF. At 330, the search and decode logic module 120 may determine whether the search for a suitable digital output code D_OUT to represent the analog input voltage V_IN is complete. In response to determining the search is complete, the search and decode logic module 120 at 340 may output the digital reference voltage code D_REF via the digital output D_OUT and output an early/late flag via flag output EL_F. If the search is not complete, the search and decode logic module 120 may return to 320 so that the comparator 120 may compare the analog input voltage V_IN to the updated reference voltage V_REF that was generated by the DAC 140 per the updated digital reference voltage code D_REF. In this regard, the search and decode logic module 130 may search for the a digital code D_REF, D_OUT that causes the DAC 140 to produce an analog reference voltage V_REF that is close to the analog input voltage V_IN.

Referring now to FIGS. 4A-4D, one manner of converting analog input voltages V_IN of 7.01*VLSB (VLSB being the

voltage corresponding to the least significant bit of the digital code V_REF, D_OUT), 15.01*VLSB, 31.01*VLSB, and 62.01*VLSB to 7-bit digital code words D_REF, D_OUT per method 300 is shown. In particular, FIGS. 4A-4D depict an exhaustive binary search used by the search and decode logic module 130 that includes seven comparisons or steps to complete each conversion. As shown, after a comparison with zero to determine whether V_IN is positive or negative, the search and decode logic module 130 in FIGS. 4A-4D uses an second analog reference voltage V_REF of 32*VLSB. From this second analog reference voltage V_REF, the search and decode logic module 130 may conduct a binary search to obtain a 7-bit digital code word D_REF, D_OUT having 1-LSB accuracy in seven steps.

FIGS. 5A-5D, illustrate another manner of converting analog input voltages V_IN of 7.01*VLSB, 15.01*VLSB, 31.01*VLSB, and 62.01*VLSB to 7-bit digital codes D_REF, D_OUT having 1-LSB accuracy per method 300. In FIGS. 5A-5D, the search and decode logic 130 selects 16*VLSB ($\frac{1}{8}$ full-scale) as the second analog reference voltage V_REF rather than 32*VLSB ($\frac{1}{4}$ full scale) as shown in FIGS. 4A-4D. The second reference voltage V_REF of 16*VLSB may have been selected based, for example, on characteristics of the analog input signal V_IN to be digitized. For example, the second reference voltage V_REF of 16*VLSB may have been selected because the probability of the analog input voltage V_IN having a magnitude less than 16*VLSB is greater than a threshold probability T, where the threshold probability T is selected based, for example, on a desired tradeoff between power consumption, speed, and/or accuracy of conversions.

It can be seen in FIGS. 5A and 5B that, where the sample to be digitized adheres to the probabilities and is in fact below the second reference voltage V_REF of 16*VLSB, then the ADC 110 completes the conversion to 1-LSB accuracy in six steps, rather than the seven steps used in FIGS. 4A and 4B. In FIG. 5C, however, the sample to be digitized is greater than second reference voltage V_REF of 16*VLSB. Consequently, the ADC 110 per method 300 requires seven steps to obtain the 7-bit digital word D_REF, D_OUT to 1-LSB accuracy, as it did in FIG. 4C. As for FIG. 5D, the sample to be digitized is also greater than 16*VLSB. However, this time the ADC 110 per method 300 requires eight steps to obtain the 7-bit digital code word D_REF, D_OUT to 1-LSB accuracy—one more step than the conversion depicted in FIGS. 4C and 5D.

FIGS. 6A-6D, illustrate another manner of converting analog input voltages V_IN of 7.01*VLSB, 15.01*VLSB, 31.01*VLSB, and 62.01*VLSB to 7-bit digital code words D_REF, D_OUT per method 300. In FIGS. 6A-6D, the search and decode logic module 130 selects 8*VLSB ($\frac{1}{16}$ full-scale) as the second analog reference voltage V_REF rather than 32*VLSB ($\frac{1}{4}$ full scale) as shown in FIGS. 4A-4D. The second reference voltage V_REF of 8*VLSB may have been selected based, for example, on characteristics of the signal to be digitized. For example, the second reference voltage V_REF of 8*VLSB may have been selected as the second reference voltage because the probability of the magnitude being less than 8*VLSB is greater than a threshold probability T, where the threshold probability T is selected based, for example, on a desired tradeoff between power consumption, speed, and/or accuracy of conversions.

It can be seen in FIG. 6A that, where the sample to be digitized adheres to the probabilities and is in fact below the second reference voltage V_REF of 8*VLSB, then the ADC

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110 may complete the conversion to 1-LSB accuracy in five steps, rather than the seven steps used in FIG. 4A. In FIG. 6B, however, the sample to be digitized is greater than the second reference voltage V_{REF} of $8 \times \text{VLSB}$. Consequently, the ADC **110** per method **300** requires seven steps to obtain the 7-bit digital code word D_{REF} , D_{OUT} to 1-LSB accuracy—the same number of steps as the conversion depicted in FIG. 4B.

In FIGS. 6C and 6D, the analog input voltage samples V_{IN} to be digitized are greater than the second reference voltage V_{REF} of $8 \times \text{VLSB}$. Moreover, because the analog input voltage V_{IN} is so far away from second reference voltage V_{REF} of $8 \times \text{VLSB}$, the ADC **110** per method **300** uses eight steps to obtain the 7-bit digital word D_{REF} , D_{OUT} to 1-LSB accuracy—one more than the conversions depicted in FIGS. 4C and 4D.

In an example implementation, the amount of time available to complete a conversion may be less than the amount of time used to achieve 1-LSB accuracy. For example, in FIGS. 4A-4D, the amount of time may be fixed at the amount of time for six comparisons or steps. Consequently, the ADC **110** may convert analog input voltages V_{IN} in the range $-16 \times \text{VLSB}$ to $16 \times \text{VLSB}$ to 1-LSB accuracy, analog input voltages V_{IN} in the ranges $-32 \times \text{VLSB}$ to $-16 \times \text{VLSB}$ and $16 \times \text{VLSB}$ to $32 \times \text{VLSB}$ to 2-LSB accuracy, and analog input voltages V_{IN} in the ranges $-64 \times \text{VLSB}$ to $-32 \times \text{VLSB}$ and $32 \times \text{VLSB}$ to $64 \times \text{VLSB}$ to 4-LSB accuracy in the allotted six comparisons or steps. Such additional quantization noise for larger magnitude input voltages V_{IN} may be tolerable in, for example, systems where the data being converted undergoes subsequent error correction (e.g., LDPC). Such additional quantization noise for larger magnitude input voltages V_{IN} may also be tolerable in, for example, communications systems where the signal-to-noise ratio (SNR) is high enough and/or where such occurrences happen infrequently enough (i.e. threshold T is high enough) such that the additional quantization noise doesn't significantly impact overall performance (e.g., bit error rate remains above a minimum threshold).

In an example implementation, the amount of time available to complete a conversion may be fixed (e.g., because downstream circuitry cannot tolerate jitter in the conversion time). Where the fixed amount of time is not the minimum conversion time, the ADC **110** may use conversion time freed-up as a result of a sample adhering to the probabilities to iterate on the LSB(s) and average the conversion results to reduce the impact of thermal noise. For example, in FIGS. 5A-5D if the conversion time were fixed at seven steps, then:

conversions of input voltages V_{IN} between $-16 \times \text{VLSB}$ to $16 \times \text{VLSB}$ may have an amount of time equal to one conversion step for iterating;

conversions of input voltage V_{IN} in the range $-32 \times \text{VLSB}$ to $-16 \times \text{VLSB}$ or in the range $16 \times \text{VLSB}$ to $32 \times \text{VLSB}$ may have just enough time to complete a conversion to 1-LSB accuracy without any iteration; and

conversions of input voltages V_{IN} in the range $-64 \times \text{VLSB}$ to $-32 \times \text{VLSB}$ or in the range $32 \times \text{VLSB}$ to $64 \times \text{VLSB}$ may only have time to complete to 2-LSB accuracy.

In other words, higher accuracy may be achieved for higher-likelihood values and less accuracy may be achieved for lower-likelihood values.

In FIGS. 6B-6D, where the sample to be digitized was greater than the second threshold, the search and decode logic module **130** at **350** jumps from the reference voltage V_{REF} of $8 \times \text{VLSB}$ back to an analog reference voltage

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V_{REF} of $32 \times \text{VLSB}$ (i.e., the voltage associated with an expected midpoint of an operating range of the analog input voltage V_{IN}). In FIGS. 7A-7C, on the other hand, the search and decode logic module **130** backs the reference voltage V_{REF} off more slowly toward the expected midpoint of an operating range of the analog input voltage V_{IN} . For example, instead of jumping directly back to the expected midpoint (e.g., $32 \times \text{VLSB}$), the search and decode logic module **130** may update the reference voltage V_{REF} from $8 \times \text{VLSB}$ to a voltage (e.g., $16 \times \text{VLSB}$) half way between the current reference voltage and the expected midpoint. Consequently, in FIG. 7A, the ADC **110** is able to complete the conversion to 1-LSB accuracy in six steps—as opposed to the seven steps depicted in FIG. 6B. In FIG. 7B, the slower back-off results in eight steps being required to achieve 1-LSB accuracy (the same number of steps as the conversion depicted in FIG. 6C). In FIG. 7C, however, the slower back-off results in an extra conversion (nine) step as compared to the conversion depicted in FIG. 6D. Thus, the faster back-off of FIGS. 6A-6D may be chosen where, for example, analog input voltages V_{IN} have a high-probability of being between $-8 \times \text{VLSB}$ and $8 \times \text{VLSB}$, a low-probability of being between $-16 \times \text{VLSB}$ and $-8 \times \text{VLSB}$ or between $8 \times \text{VLSB}$ and $-16 \times \text{VLSB}$, and a medium-probability of being between $-64 \times \text{VLSB}$ and $-16 \times \text{VLSB}$ or between $16 \times \text{VLSB}$ and $64 \times \text{VLSB}$. Conversely, the slower back-off of FIGS. 7A-7C may be chosen where, for example, analog input voltages V_{IN} having a high-probability of being between $-8 \times \text{VLSB}$ and $8 \times \text{VLSB}$, a medium-probability of being between $-16 \times \text{VLSB}$ and $-8 \times \text{VLSB}$ or between $8 \times \text{VLSB}$ to $-16 \times \text{VLSB}$, and a low-probability of being between $-64 \times \text{VLSB}$ and $-16 \times \text{VLSB}$ or between $16 \times \text{VLSB}$ and $64 \times \text{VLSB}$.

Although the searches in each of FIGS. 4A-7C begin with a comparison to zero, in other implementations the initial D_{REF} may be a value other than zero. For example, in FIGS. 5A-5D, if the probability of V_{IN} being a positive signal is above a threshold, then $16 \times \text{VLSB}$ may be the initial V_{REF} rather than the second value of V_{REF} . In such an implementation, FIG. 5B completes to 1-LSB accuracy in five steps, FIG. 5C completes to 1-LSB accuracy in six steps, and FIG. 5D completes to 1-LSB accuracy in seven steps. If V_{IN} were in fact negative, however, completion to 1-LSB accuracy may require more steps than a conventional binary search.

In an example implementation, during a conversion, the search and decode logic module **130** at **350** may select a next reference voltage V_{REF} based on an indication of how far above or below the analog input voltage V_{IN} is from the current reference voltage V_{REF} . For example, the comparator **120** may take a long time to settle or have a slow voltage slope with little or no overshoot. Such behavior of the comparator **120** may indicate that the current reference voltage V_{REF} is close to the input voltage V_{IN} and the next reference voltage V_{REF} may therefore be selected to be close to the current reference voltage V_{REF} . On the other hand, if the comparator **120** settles very quickly or has a steep voltage slope with a lot of overshoot, then the search and decode logic module **130** may select at **350** the next reference voltage V_{REF} to be far from the current reference voltage V_{REF} . For example, the search and decode logic module **130** may select the fast back-off (FIG. 6D) for a 7-bit conversion of an analog input voltage V_{IN} of 62.01 in response to the comparator **120** rapidly settling when comparing analog input voltage V_{IN} of 62.01 and to the second reference voltage V_{REF} of $8 \times \text{VLSB}$. On the other hand, the search and decode logic module **130** may select the

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slow back-off (FIG. 7A) for a 7-bit conversion of an analog input voltage V_{IN} of 15.01 in response to the comparator **120** slowly settling when comparing the input voltage V_{IN} of 15.01 and the reference voltage V_{REF} of $8 \times \text{VLSB}$.

In an example implementation, downstream circuitry may be operable to tolerate jitter in conversion time. For example, the downstream circuitry may comprise a variable rate interpolator which may interpolate the jitter conversions to fixed rate samples. The interpolator may use, for example, one or more flag signals (e.g., “early/late flag” of FIG. 2) that indicate whether the conversion completed “1 step early,” “1 step late,” “2 steps early” and so on.

In an example implementation, the search sequence may be adapted based on the PAPR of the analog input signal V_{IN} to be digitized. For example, where the input signal V_{IN} has a PAPR above a first threshold, the search and decode logic module **130** may use a first search sequence (e.g., having a second reference voltage of $16 \times \text{VLSB}$) and where input signal V_{IN} has a PAPR above a second threshold that is greater than the first threshold, the search and decode logic module **130** at **350** may use a second search sequence (e.g., having a second reference voltage of $8 \times \text{VLSB}$).

In an example implementation, gain (positive or negative) applied to the analog input voltage signal V_{IN} to be digitized may be controlled in a manner coordinated with the selection/adaptation of the search sequence.

Other embodiments of the invention may provide a non-transitory computer readable medium and/or storage medium, and/or a non-transitory machine readable medium and/or storage medium, having stored thereon, a machine code and/or a computer program having at least one code section executable by a machine and/or a computer, thereby causing the machine and/or computer to perform the processes as described herein.

Accordingly, the present invention may be realized in hardware, software, or a combination of hardware and software. The present invention may be realized in a centralized fashion in at least one computing system, or in a distributed fashion where different elements are spread across several interconnected computing systems. Any kind of computing system or other apparatus adapted for carrying out the methods described herein is suited. A typical combination of hardware and software may be a general-purpose computing system with a program or other code that, when being loaded and executed, controls the computing system such that it carries out the methods described herein. Another typical implementation may comprise an application specific integrated circuit or chip.

The present invention may also be embedded in a computer program product, which comprises all the features enabling the implementation of the methods described herein, and which when loaded in a computer system is able to carry out these methods. Computer program in the present context means any expression, in any language, code or notation, of a set of instructions intended to cause a system having an information processing capability to perform a particular function either directly or after either or both of the following: a) conversion to another language, code or notation; b) reproduction in a different material form.

While the present invention has been described with reference to certain embodiments, it will be understood by those skilled in the art that various changes may be made and equivalents may be substituted without departing from the scope of the present invention. In addition, many modifications may be made to adapt a particular situation or material to the teachings of the present invention without departing

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from its scope. Therefore, it is intended that the present invention not be limited to the particular embodiment disclosed, but that the present invention will include all embodiments falling within the scope of the appended claims.

What is claimed is:

1. A method of an analog-to-digital converter, comprising: selecting, by the analog-to-digital converter, a reference voltage;
- updating, by the analog-to-digital converter, the reference voltage based on an indication of how far an analog input voltage is from the reference voltage; and repeating, one or more times by the analog-to-digital converter, said updating to obtain a digital output code representative of the analog input voltage.
2. The method of claim 1, further comprising determining the indication of how far the analog input voltage is from the reference voltage based upon settling behavior of a comparator that is configured to compare the reference voltage to the analog input voltage.
3. The method of claim 1, further comprising: comparing, via a comparator of the analog-to-digital converter, the analog input voltage to the reference voltage;
- determining that the analog input voltage is close to the reference voltage in response to the comparator having a first settling time during said comparing; and determining that the analog input voltage is far from the reference voltage in response to the comparator having a second settling time during said comparing; wherein the second settling time is shorter than the first settling time.
4. The method of claim 1, further comprising: comparing, via a comparator of the analog-to-digital converter, the analog input voltage to the reference voltage;
- determining that the analog input voltage is close to the reference voltage in response to the comparator having a first voltage slope time during said comparing; and determining that the analog input voltage is far from the reference voltage in response to the comparator having a second voltage slope during said comparing; wherein the second voltage slope is greater than the first voltage slope.
5. The method of claim 1, further comprising: comparing, via a comparator of the analog-to-digital converter, the analog input voltage to the reference voltage;
- determining that the analog input voltage is close to the reference voltage in response to the comparator having a first overshoot during said comparing; and determining that the analog input voltage is far from the reference voltage in response to the comparator having a second overshoot during said comparing; wherein the second overshoot is greater than the first overshoot.
6. The method of claim 1, wherein said updating comprises selecting between a faster back-off process and a slower back-off process of the reference voltage based on the indication of how far the analog input voltage is from the reference voltage.
7. The method of claim 1, wherein said updating comprises: selecting a first back-off process that updates the reference voltage at a first rate in response to determining that the analog input voltage is a first distance from the reference voltage; and

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selecting a second back-off process that updates the reference voltage at a second rate in response to determining that the analog input voltage is a second distance from the reference voltage;

wherein the first distance is less than the second distance; 5
and

wherein the first rate is less than the second rate.

8. The method of claim 1, further comprising outputting a flag signal that indicates whether the digital output code was obtained early with respect to a baseline number of comparisons. 10

9. The method of claim 1, further comprising outputting a flag signal that indicates whether the digital output code was obtained late with respect to a baseline number of comparisons. 15

10. The method of claim 1, wherein said selecting is based on peak-to-average power ratio of the analog input voltage.

11. A successive approximation analog-to-digital converter, comprising:

a digital-to-analog converter operable to generate a reference voltage based on a digital reference code; 20

a comparator operable to compare an analog input voltage to the reference voltage generated by the digital-to-analog converter and generate a comparison output indicative of the comparison; and 25

a search and decode logic module comprising one or more circuits operable to:

select the digital reference code;

update the digital reference code based on the comparison output and an indication of how far the analog input voltage is from the reference voltage; 30
and

repeat, one or more times, said update of the digital reference code to obtain a digital output code representative of the analog input voltage. 35

12. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to determine the indication of how far the analog input voltage is from the reference voltage based upon settling behavior of the comparator. 40

13. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to:

determine that the analog input voltage is close to the reference voltage in response to the comparator having a first settling time; and 45

determine that the analog input voltage is far from the reference voltage in response to the comparator having a second settling time; 50

wherein the second settling time is shorter than the first settling time.

14. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to: 55

determine that the analog input voltage is close to the reference voltage in response to the comparator having a first voltage slope time; and

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determine that the analog input voltage is far from the reference voltage in response to the comparator having a second voltage slope;

wherein the second voltage slope is greater than the first voltage slope.

15. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to:

determine that the analog input voltage is close to the reference voltage in response to the comparator having a first overshoot; and

determine that the analog input voltage is far from the reference voltage in response to the comparator having a second overshoot;

wherein the second overshoot is greater than the first overshoot.

16. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to select between a faster back-off process and a slower back-off process of the reference voltage based on the indication of how far the analog input voltage is from the reference voltage.

17. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to:

select a first back-off process that updates the reference voltage at a first rate in response to determining that the analog input voltage is a first distance from the reference voltage; and

select a second back-off process that updates the reference voltage at a second rate in response to determining that the analog input voltage is a second distance from the reference voltage;

wherein the first distance is less than the second distance; and

wherein the first rate is less than the second rate.

18. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to output a flag signal that indicates whether the digital output code was obtained early with respect to a baseline number of comparisons.

19. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to output a flag signal that indicates whether the digital output code was obtained late with respect to a baseline number of comparisons.

20. The successive approximation analog-to-digital converter of claim 11, wherein the one or more circuits of the search and decode logic module are further operable to select the digital reference code based on peak-to-average power ratio of the analog input voltage.

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